

HIGH VOLTAGE POWER OPERATIONAL AMPLIFIERS

PA95

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FEATURES

- HIGH VOLTAGE 900V (±450V)
- LOW QUIESCENT CURRENT 1.6mA
- HIGH OUTPUT CURRENT 100mA
- PROGRAMMABLE CURRENT LIMIT

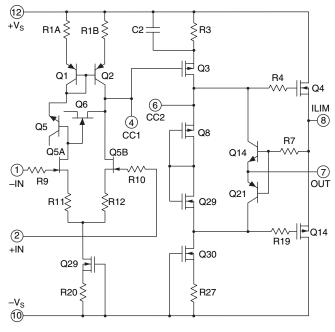
APPLICATIONS

- HIGH VOLTAGE INSTRUMENTATION
- PROGRAMMABLE POWER SUPPLIES UP TO ±430V
- MASS SPECTROMETERS
- SEMICONDUCTOR MEASUREMENT EQUIPMENT

DESCRIPTION

The PA95 is a high voltage, MOSFET operational amplifier designed as a low cost solution for driving continuous output currents up to 100mA and pulse currents up to 200mA into capacitive loads. The safe operating area (SOA) has no second breakdown limitations and can be observed for all load types by choosing an appropriate current limiting resistor. The MOSFET output stage is biased AB for linear operation. External compensation provides flexibility in choosing bandwidth and slew rate for the application. APEX's Power SIP package uses a minimum of board space allowing for high density circuit boards. The Power SIP package is electrically isolated. Isolating thermal washers (TW13) are recommended to prevent arcing from pins to heatsink.

EQUIVALENT SCHEMATIC





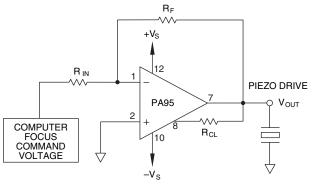
PATENTED

8-PIN SIP PACKAGE STYLE DO

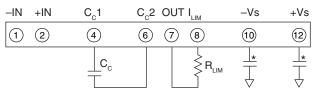
Formed leads available See package styles EC & EP

TYPICAL APPLICATION

Piezo positioning may be applied to the focusing of segmented mirror systems. The composite mirror may be composed of hundreds of elements, each requiring focusing under computer control. In such complex systems the PA95 reduces the costs of power supplies and cooling with its advantages of low cost and low quiescent power consumption while increasing circuit density with the SIP package.



EXTERNAL CONNECTIONS



* .01F or greater ceramic power supply bypassing required.

PHASE COMPENSATION

≥100

CC rated for full supply voltage.

$$R_{\text{LIM}} = \frac{I}{I_{\text{LIM}}}$$

ABSOLUTE MAXIMUM RATINGS

SUPPLY VOLTAGE, +V $_{\rm S}$ to –V $_{\rm S}$ OUTPUT CURRENT, source, sink 900V 200mA, within SOA POWER DISSIPATION, continuous @ $T_c = 25^{\circ}C$ 30W INPUT VOLTAGE, differential ±20V INPUT VOLTAGE, common mode³ ±۷。 260°C TEMPERATURE, pin solder - 10s max. 150°C TEMPERATURE, junction² TEMPERATURE RANGE, storage -40 to +85°C OPERATING TEMPERATURE RANGE, case -25 to +85°C

SPECIFICATIONS

PARAMETER	TEST CONDITIONS ¹	MIN	ТҮР	MAX	UNITS
INPUT OFFSET VOLTAGE, initial OFFSET VOLTAGE, vs. temperature OFFSET VOLTAGE, vs. supply OFFSET VOLTAGE, vs. time BIAS CURRENT, initial BIAS CURRENT, vs. supply OFFSET CURRENT, initial INPUT IMPEDANCE, DC INPUT CAPACITANCE COMMON MODE VOLTAGE RANGE ³ COMMON MODE REJECTION, DC	Full temperature range Vs=±250V SEE NOTE 3 V _{CM} = ±90V	±V _s ∓30 80	.5 15 10 75 200 4 50 10 ¹¹ 4 98	5 50 25 2000 500	mV μV/°C μV/V μV/kh pA pA/V pA Ω pF V dB
NOISE GAIN OPEN LOOP, @ 15Hz GAIN BANDWIDTH PRODUCT at 1MHz POWER BANDWIDTH PHASE MARGIN,Av=10	10KHz BW, $R_s = 1KΩ$ $R_L = 5KΩ$ $R_L = 5KΩ$ $R_L = 5KΩ$ Full temperature range	94	2 118 10 20 60		µVrms dB MHz kHz ∘
OUTPUT VOLTAGE SWING CURRENT, continuous SLEW RATE, $A_v = 100$ SETTLING TIME to .1% RESISTANCE	$I_o = 70$ mA $C_c = 4.7$ pF 2V step no load	±V _s ∓24 100	±V _s ∓20 30 1 100		V mA V/μs μs Ω
POWER SUPPLY VOLTAGE⁵ CURRENT, quiescent	See note 5	±50	±300 1.6	±450 2.2	V mA
THERMAL RESISTANCE, AC, junction to case ⁴ RESISTANCE, DC, junction to case RESISTANCE, junction to air TEMPERATURE RANGE, case	Full temperature range, F > 60Hz Full temperature range, F < 60Hz Full temperature range	-25	30	2.5 4.2 +85	°C/W °C/W °C/W °C

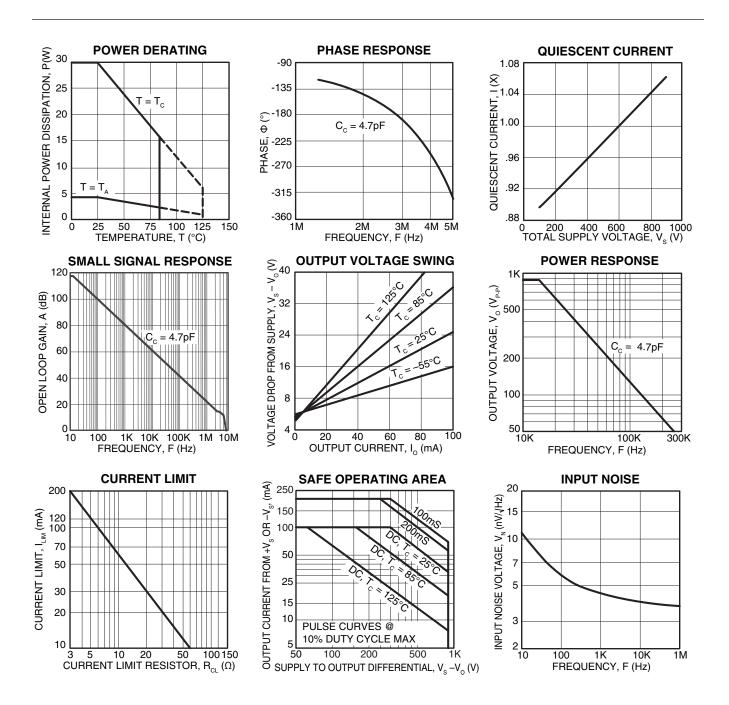
NOTES: 1. Unless otherwise noted: $T_c = 25^{\circ}$ C, DC input specifications are ± value given. Power supply voltage is typical rating. $C_c = 4.7$ pF. 2. Long term operation at the maximum junction temperature will result in reduced product life. Derate internal power dissipation to

- achieve high MTTF.
- Although supply voltages can range up to ± 450V the input pins cannot swing over this range. The input pins must be at least 30V from either supply rail but not more than 500V from either supply rail. See text for a more complete description of the common mode voltage range.
- 4. Rating applies if the output current alternates between both output transistors at a rate faster than 60Hz.
- 5. Derate max supply rating .625 V/°C below 25°C case. No derating needed above 25°C case.

CAUTION

The PA95 is constructed from MOSFET transistors. ESD handling procedures must be observed.

The exposed substrate contains beryllia (BeO). Do not crush, machine, or subject to temperatures in excess of 850°C to avoid generating toxic fumes.



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GENERAL

Please read Application Note 1 "General Operating Considerations" which covers stability, supplies, heat sinking, mounting, current limit, SOA interpretation, and specification interpretation. Visit www.apexmicrotech.com for design tools that help automate tasks such as calculations for stability, internal power dissipation, current limit; heat sink selection; Apex's complete Application Notes library; Technical Seminar Workbook; and Evaluation Kits.

CURRENT LIMIT

For proper operation, the current limit resistor $(\mathrm{R}_{_{\mathrm{LIM}}})$ must be connected as shown in the external connection diagram. The minimum value is 3.5 ohm, however for optimum reliability the resistor value should be set as high as possible. The value is calculated as follows; with the maximum practical value of 150 ohms.

$$R_{LIM} = \frac{.7}{I_{LIM}}$$

COMMON MODE INPUT RANGE

Operational amplifiers are usually designed to have a common mode input voltage range that approximates the power supply voltage range. However, to keep the cost as low as possible and still meet the requirements of most applications the common mode input voltage range of the PA95 is restricted. The input pins must always be a least 30V from either supply voltage but never more than 500V. This means that the PA95 cannot be used in applications where the supply voltages are extremely unbalanced. For example, supply voltages of +800V and -100V would not be allowed in an application where the non-inverting pin is grounded because in normal operation both input pins would be at 0V and the difference voltage between the positive supply and the input pins would be 800V. In this kind of application, however, supply voltages +500V and -100V does meet the input common mode voltage range requirements since the maximum difference voltage between the inputs pins and the supply voltage is 500V (the maximum allowed). The output has no such restrictions on its voltage swing. The output can swing within 24V of either supply voltage regardless of value so long as the total supply voltage does not exceed 900V.

INPUT PROTECTION

Although the PA95 can withstand differential input voltages up to ±20V, additional external protection is recommended. In most applications 1N4148 or 1N914 signal diodes are sufficient (D1, D2 in Figure 1a). In more demanding applications where low leakage or low capacitance are of concern 2N4416 or 2N5457-2N5459 JFETs connected as diodes will be required (Q1, Q2 in Figure 1b). In either case the input differential voltage will be clamped to ±.7V. This is sufficient overdrive to produce maximum power bandwidth. Note that this protection does not automatically protect the amplifier from excessive common mode input voltages.

POWER SUPPLY PROTECTION

Unidirectional zener diode transient suppressors are recommended as protection on the supply pins. The zeners clamp transients to voltages within the power supply rating and also

clamp power supply reversals to ground. Whether the zeners are used or not, the system power supply should be evaluated for transient performance including power-on overshoot and power-off polarity reversal as well as line regulation.

Conditions which can cause open circuits or polarity reversals on either power supply rail should be avoided or protected against. Reversals or opens on the negative supply rail is known to induce input stage failure. Unidirectional transzorbs prevent this, and it is desirable that they be both electrically and physically as close to the amplifier as possible.

STABILITY

The PA95 is stable at gains of 10 or more with a NPO (COG) compensation capacitor of 4.7pF. The compensation capacitor, Cc, in the external connections diagram must be rated at 1000V working voltage and mounted closely to pins 4 and 6 to prevent spurious oscillation. A compensation capacitor less than 4.7pF is not recommended.

EXTERNAL COMPONENTS

The compensation capacitor Cc must be rated for the total supply voltage. An NPO (COG)capacitor rated a 1kV is recommended.

Of equal importance are the voltage rating and voltage coefficient of the gain setting feedback resistor. Typical voltage ratings of low wattage resistors are 150 to 250V. Up to 500 V can appear across the feedback resistor. High voltage rated resistors can be obtained. However a 1 megohm feedback resistor composed of five 200k resistors in series will produce the proper voltage rating.

CAUTIONS

The operating voltages of the PA95 are potentially lethal. During circuit design develop a functioning circuit at the lowest possible voltages. Clip test leads should be used for "hands off" measurements while troubleshooting.

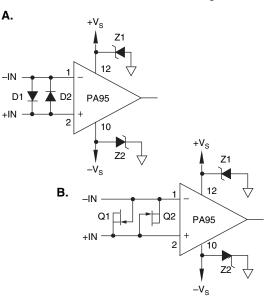


FIGURE 1. OVERVOLTAGE PROTECTION

This data sheet has been carefully checked and is believed to be reliable, however, no responsibility is assumed for possible inaccuracies or omissions. All specifications are subject to change without notice. PA95U REV H NOVEMBER 2004 © 2004 Apex Microtechnology Corp. 4